

Fig. 4—Parameters of an inductive post in WR-187 waveguide.

where  $w$  and  $l$  are the amplitude and period shown in Fig. 3. Thus,

$$k = \tan \frac{\pi}{4} \left( 1 - \frac{2w}{l} \right). \quad (16)$$

Eq. (16) is a very practical expression, because plotting errors can be averaged out by using the ratio of the average amplitude  $\bar{w}$  to the average period  $\bar{l}$ .

The intrinsic angular lengths,  $\theta_1 = \beta_1 d_1$  and  $\theta_2 = \beta_2 d_2$ , are given by

$$\theta_1 = \beta_1 D_0 - \frac{\pi}{4} \left( 1 - \frac{2w}{l} \right) \quad (17)$$

$$\theta_2 = \beta_2 S_0 - \frac{\pi}{4} \left( 1 - \frac{2w}{l} \right) \quad (18)$$

where  $\beta_1 D_0$  and  $\beta_2 S_0$  are the coordinates of an "inside peak," such as point ②. The correct point to use is the one having the most nearly equal values of  $\beta_1 D$  and  $\beta_2 S$ . If the network happens to have bilateral symmetry about plane  $T_0$ ,  $\beta_1 D_0$  and  $\beta_2 S_0$  will be exactly equal; in general, they will differ by less than  $\pi/2$ .

An application for this measurement arises in the design of filters using thick inductive posts. The impedance behavior with frequency of a thick post does not follow that of a narrow iris, and the intrinsic angular length cannot be computed from the handbook value of the shunt susceptance. In Fig. 4 are shown values of  $\theta$  and  $k$  derived from tangent-relation plots of data taken with centered inductive posts in WR-187 waveguide at 5.5 Gc. It is interesting that the sign of  $\theta$  changes at a diameter of approximately  $a/6$ . For posts of larger diameter,  $d_1 = d_2 = \theta/2\beta$  is positive, and directly-coupled resonator sections (of a cascade-resonator filter) must be slightly longer than  $\lambda_{\theta_0}/2$ .

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## Integral Quotient in Measurements of Ambipolar Diffused Plasma with TE<sub>011</sub> Cavity

Despite some known exact solutions of plasma loaded TM<sub>010</sub> cavity,<sup>1,2</sup> the TE<sub>011</sub> mode should be used due to measurement-technical reasons in the case of large electron densities and considerable losses.<sup>3,4</sup> This communication is connected with the mathematical treatment of measurement results in the case of TE<sub>011</sub> mode and ambipolar diffusion in discharge tube. The treatment is based on Slater's<sup>5</sup> expression for discharge admittance and the assumption of small perturbations, but not on any special electron theoretically derived plasma conductivity formula.

Owing to the ambipolar diffusion in the discharge tube, the distribution of electron density  $n$  along the radius  $r$  is

$$n = n_0 J_0(2.405r/r_0) \quad (1)$$

where  $n_0$  is the electron density at the axis and  $r_0$  the inside radius of the discharge tube. The plasma conductivity is proportional to the electron density and obeys the same distribution.

Starting from Slater's discharge admittance,<sup>5</sup> the expressions for the real part  $\sigma_{r0}$  and the imaginary part  $\sigma_{i0}$  of plasma conductivity at the axis can be derived and are

$$\begin{cases} \sigma_{r0} = (g_d \omega_0 \epsilon_0 / \beta) Q \\ \sigma_{i0} = (-2\Delta\omega_0 \epsilon_0) Q \end{cases} \quad (2)$$

where  $g_d$  is the discharge conductance,  $\omega_0$  the angular resonant frequency,  $\Delta\omega_0$  the change of  $\omega_0$  due to the discharge plasma,  $\epsilon_0$  the dielectric constant of free space,  $\beta$  the factor depending on coupling between transmission line and cavity, and  $Q$  the quotient of two volume integrals.

In the case of ambipolar diffusion and TE<sub>011</sub> cavity, after integration with respect to  $z$  and  $\phi$  (in cylindrical coordinates)<sup>6</sup> one has

$$Q = Q(R/r_0) = Q_1(R)/Q_2(R/r_0, r_0) = \int_0^R r J_1^2(3.832r/R) dr / \int_0^{r_0} r J_0(2.405r/r_0) J_1^2(3.832r/R) dr. \quad (3)$$

$R$  is the inner radius of the cavity.

The integration in the numerator of (3) can be performed simply and one has  $Q_1(R) = \frac{1}{2} R^2 J_0^2(3.832)$ . The values of  $Q_2(R/r_0, r_0)$  at  $r_0 = 1$  cm have been computed on an electronic digital computer. The results are presented in Table I.  $Q_2$  can be found for

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<sup>1</sup> B. Agdur and B. Enander, "Resonances of a microwave cavity partially filled with a plasma," *J. Appl. Phys.*, vol. 33, pp. 575-581; February, 1962.

<sup>2</sup> P. Hedvall, "Cavity method for measuring plasma properties," *Ericsson Technics*, vol. 19, pp. 97-107; 1963.

<sup>3</sup> K. B. Persson, "Limitations of the microwave cavity method of measuring electron densities in a plasma," *Phys. Rev.*, vol. 106, pp. 191-195; April, 1957.

<sup>4</sup> S. J. Buchsbaum and S. C. Brown, "Microwave measurements of high electron densities," *Phys. Rev.*, vol. 106, pp. 196-199; April, 1957.

<sup>5</sup> J. C. Slater, "Microwave electronics," *Rev. Mod. Phys.*, vol. 18, pp. 441-512; October, 1946.

<sup>6</sup> P. Jaaskeläinen, "On attenuation and electrical length of a plasma loaded helical transmission line," *Acta Polytech. Scand.*, Ph 23; 1963.

TABLE I  
COMPUTED VALUES OF DENOMINATOR  $Q_2$  AT  $r_0 = 1$  CM  
AND INTEGRAL QUOTIENT  $Q$  AS FUNCTION OF  $R/r_0$ ,  
THE RATIO OF CAVITY RADIUS TO  
DISCHARGE TUBE RADIUS

$R/r_0$	$Q_2, \text{cm}^2$	$Q$
3	0.022358	32.650
4	0.013688	94.805
5	0.0091115	222.54
6	0.0064640	451.71
7	0.0048106	826.14
8	0.0037141	1397.6
9	0.0029514	2225.9
10	0.0024005	3378.8

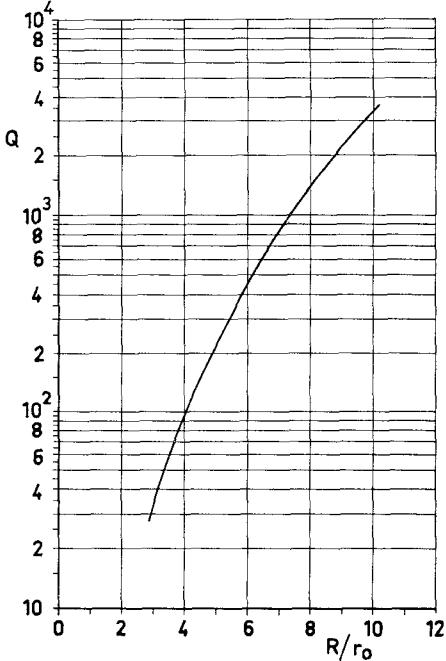


Fig. 1—The quotient  $Q$  of two volume integrals in the measurement of ambipolar diffused plasma with TE<sub>011</sub> cavity.  $R$  is the inner radius of cavity,  $r_0$  the inner radius of discharge tube.

other values of  $r_0$  by noting that it is proportional to the square of  $r_0$ . The integral quotient  $Q$  is only the function of  $R/r_0$  and is presented in Table I and in Fig. 1 for practical values of the argument  $R/r_0$ .

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## On "Status Report on International Millimeter Waveguide Flange Standards"<sup>1</sup>

In his communication, Anderson described the state of international standardization of millimeter waveguide flanges and concluded that in the absence of a suitable

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<sup>1</sup> T. N. Anderson, IEEE TRANS. ON MICROWAVE THEORY AND TECHNIQUES, vol. MTT-11, pp. 427-429; September, 1963.